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## Grid Side Inverter Control for a Grid Connected Synchronous Generator Based Wind Turbine Experimental Emulator

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### ABSTRACT

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#### Keywords:

*wind energy conversion system, synchronous generator, grid side converter, DC bus voltage, dSPACE DS1104*

This paper describes the real time implementation and control of a wind energy conversion chain emulator based on a synchronous generator (SG) using a full-scale power converter configuration. The proposed structure consists of the mechanical coupling of two 1.5 kW machines, a DC motor which emulates the static-dynamic behaviors of a three-blade wind turbine with a horizontal axis including an ideal gearbox, and a synchronous generator that ensures the electromechanical conversion and manages the different operating modes. The aim of the first part in this work is the design and the implementation of the control of the grid side converter in order to control the flow of the produced/consumed active and reactive power ( $P_{GSC} / Q_{GSC}$ ) in both directions between the generator and the grid. An improved experimental grid voltage vector-orientation control algorithm (VOC) is investigated and applied to the grid inverter to control the GSC powers independently and instantly. The control algorithms are implanted in C, using dSPACE DS1104 control board to drive the 6-IGBT's inverter. The experimental results validate the effectiveness of the proposed control scheme of the GSC.

## 1. INTRODUCTION

Research and teaching interests in renewable energies and generation systems are growing. The projects at the laboratory concerns the design, implantation and intelligent management of an energy production system which is composed of two wind turbine emulator (double fed induction generator (DFIG) and synchronous generator (SG)), and a photovoltaic system. The results of this work will be used for training, PhD researchers and also for national needs in the field of renewable energies [1, 2].

This work is a continuation of a previous theoretical and practical research in the field of wind energy conversion system, that falls with the lab framework project that aims to the real-time design, implementation and intelligent management of smart distributed renewable energy micro-power plants [3, 4]. The project develops four types of micro-power plants: 1.5 kW Double Fed Induction Generator Based Wind Turbine Emulator, 1.5 kW Synchronous Generator Based Wind Turbine Emulator, Photovoltaic system with Grid Tie Inverters and finally, a 250 W Photovoltaic micro-power plant connected through a self-made single phase Grid Tie Inverter [2, 5, 6].

Furthermore, as first part, our previous works carried out in papers [1, 2], focuses on the experimental implementation of a grid connected variable-speed concept based wind power system, with partly rated power converters. The emulator is equipped with the DFIG configuration. The aim of this research is to build a low-cost experimental bench for the emulation of a wind turbine's conversion chain, with the use of a controlled DC armature motor for the aerodynamic drive emulation. In both hypo synchronous and hypersynchronous

modes. This device handles multiple scenarios and operating points, and allow to evaluate numerous control laws (linear / non-linear) for the generator power control. The power HIL was tested successfully thanks to the DFIG PI linear control law. However, a nonlinear control law would be more flexible in order to improve the quality of the control and reduce the operating constraints.

The second prototype consists of a 1.5 kW photovoltaic industrial micro power plant, six 250 W PV panels connected to three Grid Tie Inverters (GTI), with power line communication Unit (ECU). A Raspberry Pi3 (RPi3) reads the data of the generated power from the web server of the ECU and send them to the Firebase Database (DB). We develop strategies to control the micro power plants in a way that each one produces the exact quantity of power in real-time to follow the consumption variation and the weather changes.

The third system is a 250 W photovoltaic self-made micro power plant and is strongly related to the previous part. It implements a single-phase GTI. The study outlines indirect active and reactive power injection using a zero-crossing detection (ZCD) method to master the synchronization of the generated voltage from the inverter and the power grid [5, 6].

The successful implementation of the contribution established in this paper will allow the extraction of emulated electrical wind energy in order to inject it into the common system grid according to the needs determined by the intelligent management of the micropower plant system. On the other hand, the control technique of the grid side inverter can be added to the DFIG setup instead of the all-diode rectifier.

Nowadays, the synchronous generator based wind turbine system is one of the most popular configurations that use the

concept of full-scale power electronic converter. This is due to its large advantage in reducing sensitivity against the grid voltage disturbance, moreover, the good operating under variable speeds thanks to the full control of the generator, consequently, achieve the maximum efficiency, with a low synchronous speed thanks to a high poles number, add to that, a protection of the generator by isolating it from the grid, unlike the DFIG configuration [7-9].

Many authors have discussed the fully rated power converters based variable speed wind turbine system. Hansen and Michalke [10] discussed the modeling and control of a full-scale back-to-back converter wind turbine with multi-pole permanent magnet synchronous generator PMSG. They focus on the grid support capability in uninterrupted operation during grid faults. The paper [11] provides a detailed and robust model for a back-to-back converter of a multi-pole PMSG wind turbine. A damping controller is implemented in order to actively damp the torsional oscillations in the drive-train to absorb the system stress. The active and reactive power is controlled thank to the grid-side converter, while the machine side converter keeps the stator voltage and the DC bus voltage constant. In the paper [12], the authors use the hybrid adaptive control method for adjusting the DC link voltage in the weak-grid mode for a connected PMSG. The control algorithm is evaluated in both simulation and experimental environments. Berkani et al. [13] investigated an improvement of the control of the DC link voltage of a dual star synchronous machine. The authors propose the sliding mode approach to drive the gride side inverter, while keeping a conventional PI controller for the machine side inverter drive. The work [14] investigates the experimental implementation of a predictive control model algorithm for a connected two-level voltage source converter. The paper details clearly the four steps of the proposed technique, the authors focus on the performance and the efficiency. The control board DS1104 used is the same as in our experimental setup.

Our paper is organized as follows: Section 2 gives the detailed description of the overall proposed structure of this wind energy emulator with a presentation of all the experimental bench elements (machines, power electronic converters, measurements...). In section 3, we present the GSC modeling, including the  $(R_f, L_f)$  filter and the DC link connection, with its control algorithm. In section 4, the experimental results show the control of the active and reactive power transfer between the grid and GSC. Finally, section 5

gives some conclusions.

## 2. HARDWARE DESCRIPTION OF THE WIND TURBINE EMULATOR

Figure 1 shows the complete diagram of our experimental bench, which consists in a wind turbine emulator based on a grid connected synchronous machine. A 1.5 kW DC motor is mechanically coupled to the SG shaft in order to simulate the wind turbine drive, hence, to get the needed aerodynamic torque. The DC motor is powered through a chopper, and its power comes from an independent DC power supply source. The SG is driven by a 6-IGBT's inverter, the machine side converter (MSC) that imposes the current references for reaching the desired SG active and reactive power through vector control [1, 2, 15].

The machine side converter (MSC) is powered by a second 6-IGBT inverter which is called grid side converter (GSC) through a DC link contains a filtering capacitor. The GSC main role is the control of direction and the power consumed or injected into the grid by the acting on the phases currents  $(I_{qf}, I_{df})$ . An autotransformer is used for testing and to adapt the grid voltage to the system.

We focus in this paper on the control of the GSC, for this, we use a dSPACE DS1104 control board to generates 6 PWM signals in order to control the 3-legs IGBT's grid side inverter. The same board will be used to control the MSC and the DC chopper, by changing the internal firmware of its slave DSP and with proper programming of its master controller (PowerPC).

We use the following ADC measures: The DC bus voltage (VDC), 2 GSC phase voltages  $(V_{sa}, V_{sb})$ , 2 GSC phase currents  $(I_{fa}, I_{fb})$ . For the MSC, we will need  $(I_{msa}, I_{msb})$  and for the DC motor torque control (wind emulator), we will measure its current  $(I_{DCM})$ . We use the LEM LA55 Hall effect sensors for the ADC current measurements and the LEM LV25-P Hall effect sensors for the voltages. Proper active filtering and signal conditioning is added before to the ADC stage.

The Control Desk™ running on the DS1104 host PC allows the visualization and interaction with the system variables in real time. The lab set up is shown in Figure 13 (See Appendix), while the hardware components are exposed in Table 1 (See Appendix). The nominal parameters of the experimental bench are given in Table 2 (See Appendix).

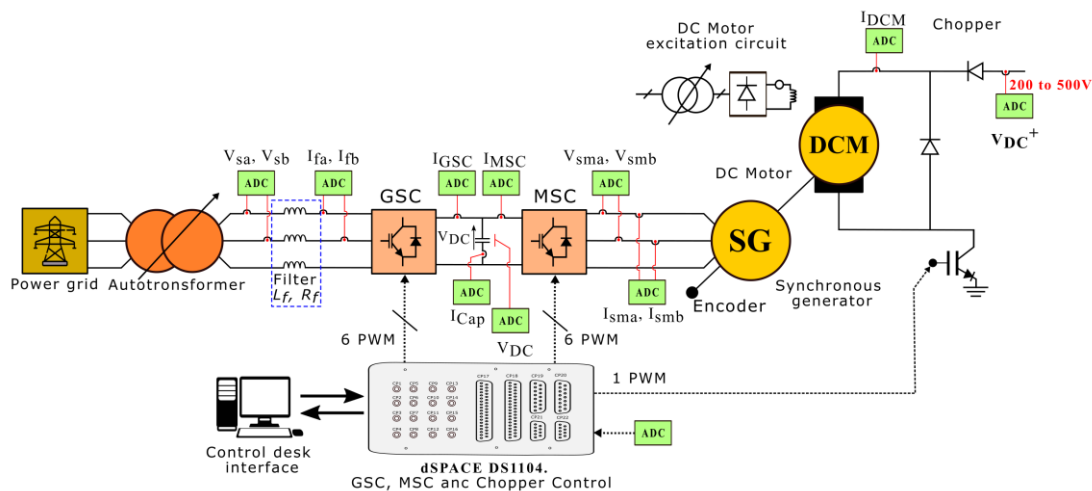


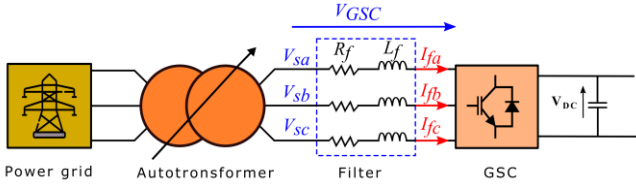
Figure 1. Experimental structure of the wind turbine emulator based on a SG

### 3. GSC MODELLING AND CONTROL

GSC is responsible for the DC bus voltage adjustment and the power flow from and to the grid. As a first step in the implementation of this emulator, we start by testing only the grid side inverter without the entire emulator, based on that, we introduce the simplified model of the filter-inverter-DC bus assembly in the next description [7].

#### 3.1 Grid-Filter-Inverter modeling

According to Kirchhoff law applied to the circuit of Figure 2, we express the phase voltages of the GSC, the grid and the filter in the  $abc$  reference framework. We consider a receptor convention [16, 17].



**Figure 2.** Representation of the GSC inverter, the filter and the grid

The phase voltages between the GSC and the grid can be expressed in the stationary  $abc$  reference frame as follows [18]:

$$V_{GSC\ abc} = R_f I_{f\ abc} + L_f \frac{dI_{f\ abc}}{dt} + V_{S\ abc} \quad (1)$$

In order to apply the vector control principle, we express the previous model in the  $dq$  reference frame. The GSC voltages ( $V_{df}$ ,  $V_{qf}$ ) are given by [8, 19]:

$$V_{df} = R_f I_{df} + L_f \frac{dI_{df}}{dt} - \omega_s L_f I_{qf} + V_{ds} \quad (2)$$

$$V_{qf} = R_f I_{qf} + L_f \frac{dI_{qf}}{dt} + \omega_s L_f I_{df} + V_{qs} \quad (3)$$

The active power and the reactive power of the GSC in the  $dq$  reference frame are described by [16]:

$$P_{GSC} = V_{ds} I_{df} + V_{qs} I_{qf} \quad (4)$$

$$Q_{GSC} = V_{qs} I_{df} - V_{ds} I_{qf} \quad (5)$$

#### 3.2 GSC control

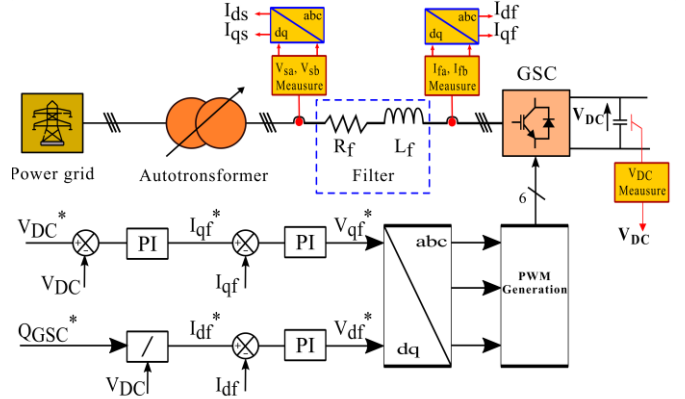
Figure 3 shows the overall structure of the GSC control, the main objective is to control the voltage in the DC link, and the active and reactive power ( $P_{GSC}$ ,  $Q_{GSC}$ ) delivered to the grid by acting on the phases current through two current control loops [7]. The control of the active power is linked to the control of the DC bus voltage [20].

To control the GSC powers  $P_{GSC}(I_{qf})$ ,  $Q_{GSC}(I_{df})$  we apply a grid voltage vector-orientation control (VOC). Thus, the grid voltage vector  $V_s$  is hold on the  $q$ -axis ( $V_{qs}=U_s$ ,  $V_{ds}=0$ ). We

give the expression of the GSC powers according to the GSC currents as follows [21].

$$P_{GSC} = U_s I_{qf} \quad (6)$$

$$Q_{GSC} = U_s I_{df} \quad (7)$$



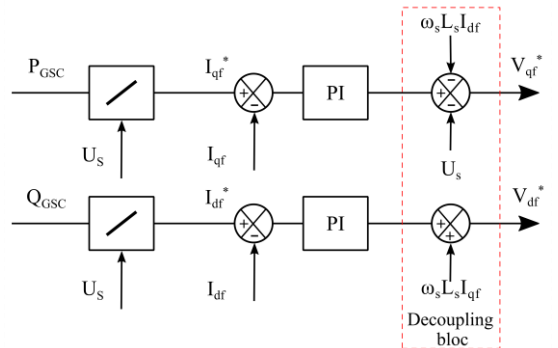
**Figure 3.** Schematic diagram of the GSC control

The decoupled model obtained allows controlling the GSC powers independently. For the GSC active power  $P_{GSC}$  control, we act on GSC quadrature current  $I_{qf}$  and we act on the GSC direct current  $I_{df}$  for GSC reactive power control [22]. The GSC voltages references, which are the outputs of the current controllers, are given by Amrane et al. [23]:

$$V_{df} = R_f I_{df} + L_f \frac{dI_{df}}{dt} - \omega_s L_f I_{qf} \quad (8)$$

$$V_{qf} = R_f I_{qf} + L_f \frac{dI_{qf}}{dt} + \omega_s L_f I_{df} + U_s \quad (9)$$

Figure 4 shows the detailed structure of the GSC control. The error signal between the measured value ( $I_{qf}$ ,  $I_{df}$ ) and the reference value ( $I_{qf}^*$ ,  $I_{df}^*$ ) are fed in a PI controller. The power reference steps ( $Q_{GSC}$  or  $I_{df}^*$ ), ( $P_{GSC}$  or  $I_{qf}^*$ ) are imposed manually through the Control Desk interface. In typical control systems, the reactive power reference is set to zero to perform the operating at a unity power factor. The introduction of the decoupling block with the correct calculation of the controller parameters ensures great performances of the power control system [24].



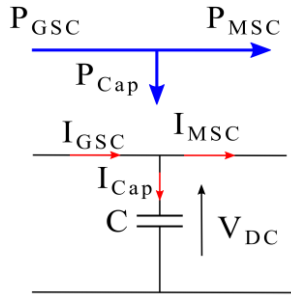
**Figure 4.** Detailed control loop diagram of the GSC currents schematic diagram of the GSC control

### 3.3 DC Bus modeling and control

The power flowed in the DC bus as shown in Figure 5 are thus expressed by [18, 25]:

$$P_{Cap} = P_{GSC} - P_{MSC} \quad (10)$$

where,  $P_{Cap}$  is the power absorbed by the capacitor,  $P_{GSC}$  the power delivered by the GSC, and  $P_{MSC}$  the power absorbed by the MSC [26]. Negative values mean that the devices are generating power [27].



**Figure 5.** DC Link circuit

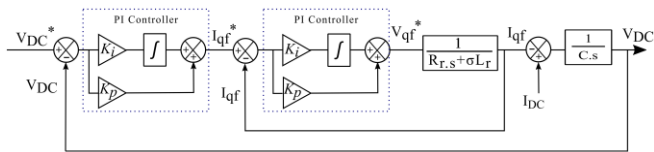
$P_{cap}$  can be expressed by its linearized expression:

$$P_{Cap} = V_{DC} C \frac{dV_{DC}}{dt} \quad (11)$$

From Eq. (10) and Eq. (11), the DC bus voltage expression is given as:

$$\frac{dV_{DC}}{dt} = \frac{1}{C} \frac{P_{GSC} - P_{MSC}}{V_{DC}} \quad (12)$$

Figure 6 shows the block diagram of the DC bus voltage control loop, the role of this control loop is keeping the DC link voltage constant and assuring that all the power coming from the MSC is instantaneously transferred to the grid through the GSC [20, 28].



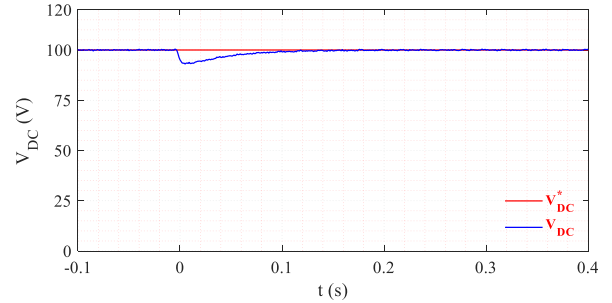
**Figure 6.** DC bus voltage control loop

## 4. EXPERIMENT RESULTS AND INTERPRETATION

Building the experimental bench is done step by step. We first evaluate the performances of the GSC control using PI controllers. Further part will be the wind turbine emulator consisting of the SG controlled by the MSC and driven by the DC motor. The real-time reference changes are made thanks to the Control Desk™ software running on the DS1104 host PC. The interface allows the visualization of the system variables and interaction in real-time. As well as the capture of data of the estimated currents, voltages and powers. Therefore, we present two tests with the Grid Side Converter control.

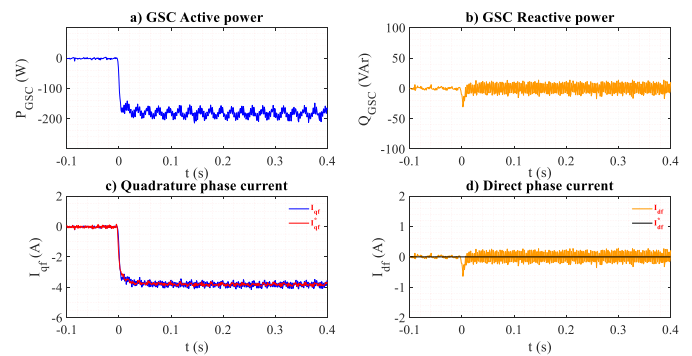
### 4.1 First test

At  $t = 0$ s, we apply a disturbance consisting in a sudden power load on the DC bus in order to show the control of the DC voltage at a reference of  $V_{DC}^* = 100$  V. The reactive power reference is scheduled at  $Q_{GSC}^* = 0$  VAR for an unity power factor. The measured DC link voltage  $V_{DC}$  perfectly follows its reference  $V_{DC}^*$  after a brief transient (Figure 7).

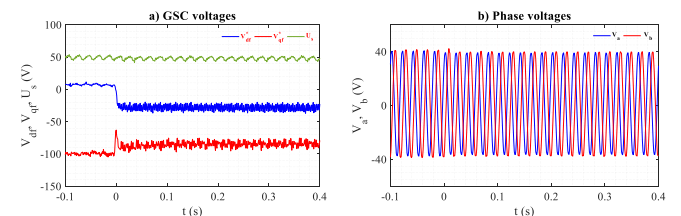


**Figure 7.** DC bus voltage control

The disturbance is rejected thanks to the injection by the control system of the right amount of active power  $P_{GSC}$  by the GSC. The DC bus link absorbs an active power from the grid. The  $V_{DC}$  control loop adjusts the reference  $I_{qf}^*$  around 4 A, in order to impose this active power. The reactive power  $Q_{GSC}$  remains constant around 0 VAR with non-important oscillations during all the test, therefore the GSC direct current reference is  $I_{df}^* = 0$  A. The estimated active  $P_{GSC}$  and reactive power  $Q_{GSC}$  fluctuations also relies on the  $U_s$  ( $V_{qs}$ ) measurement, which is not so perfect, consequently the estimated active power is around -180 W with oscillations  $\pm 20$  W (11%) max in steady state (Figures 8a and 8b). The measured GSC currents ( $I_{qf}$ ,  $I_{df}$ ) follow their references ( $I_{qf}^*$ ,  $I_{df}^*$ ) correctly thanks to the high performance of the PI controllers (Figures 8c and 8d).



**Figure 8.** GSC Power control



**Figure 9.** a) GSC voltages, b) Phase voltages

Figure 9a shows the necessary GSC voltages references ( $V_{qf}^*$ ,  $V_{df}^*$ ) needed to track the desired currents ( $I_{qf}$ ,  $I_{df}$ ) and



consequently, the power references. The grid voltage, after the autotransformer,  $U_s$  ( $V_{qs}$ ) ( $V_{qs} = 0$ ) is measured around 50 V. We used a low voltage for the test to operate with a DC voltage around 100 V. The sinusoidal voltages of the phases after the autotransformer have an amplitude of 40 V (Figure 9b).

#### 4.2 Second test

In the second test, we still apply the reference of  $V_{DC}^* = 100$  V, with a step on the reactive power  $Q_{GSC}^*$  from 200 VAR to -200 VAR. The measured DC bus voltage  $V_{DC}$  track its reference  $V_{DC}^*$  of 100 V without almost any disturbance (Figure 10).

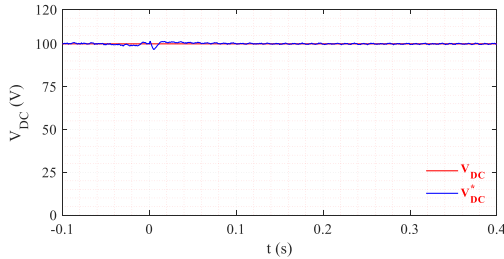


Figure 10. DC bus voltage control

The reactive power follows its reference step correctly from 200 VAR to -200 VAR thanks to the direct current control (Figure 11b). Figure 11c and 11d present the same dynamic performances of the GSC current control as the previous experiment. The curve of the controlled GSC direct current  $I_{df}$  shows a good tracking accuracy, with an -8 A step. The direct current value will be stabilized around -4 A with negligible oscillations in steady state.  $I_{df}$  response shows a single spike of 2 A in transient state, in order to reject any  $V_{DC}$  disturbances, then remains at around -4 A, delivering an active power of -200 W to the grid (Figure 11a).

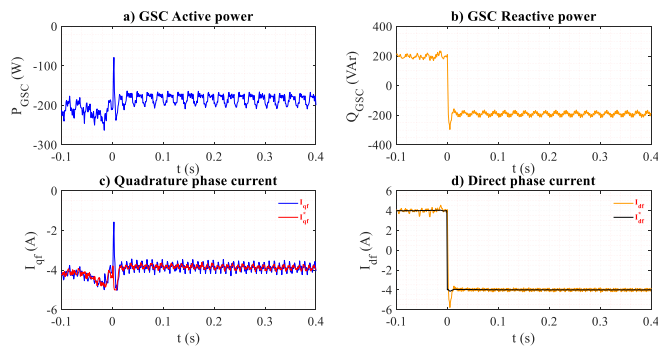


Figure 11. GSC Power control

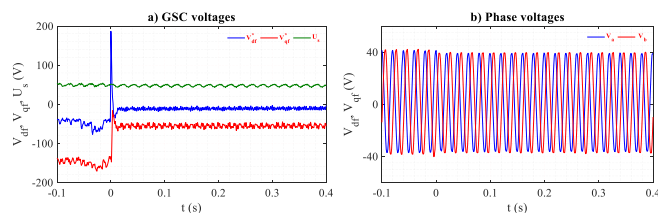


Figure 12. a) GSC Voltages, b) Phase voltages

Figure 12a shows the necessary GSC voltage references that are outputted by the ( $I_{qf}$ ,  $I_{df}$ ) PI current controllers with significant peaks at the transient state, especially  $V_{df}^*$  voltage reference, that is affected by the large reference change in  $Q_{GSC}^*$  (from 200 VAR to -200 VAR at  $t=0s$ ). The grid voltage

$U_s$  ( $V_{qs}$ ) is always constant at 50 V, (Figure 12b).

## 5. CONCLUSIONS

The study discussed in this work is part of recent research on the introduction of intelligent micro-power plants in micro-grids. A 1.5 kW synchronous generator-based wind power emulator is built according to the variable speed concept with full power electronic converter configuration. The purpose of the current work aims with the design and the control implementation of the VOC control technique, in order to drive the 6-IGBT's grid inverter. The technique allows the control of the delivered active and reactive power from and into the grid, in an instantaneous and independent way.

The grid active and reactive power are adjusted by acting on the dq grid currents, via both current control loops using PI controllers. They drive the current with enough precision and high performances. The future work is to complete this experimental setup by the implantation of the DC motor control to elaborate complex wind speed profiles in parallel with the implementation of the machine side converter (MSC) control to drive the SG. This will deliver the optimum extracted power, using MPPT, to the grid.

## ACKNOWLEDGMENT

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## NOMENCLATURE

$V_{GSCabc}$	a, b and c phase voltages of the GSC
$I_{GSCabc}$	a, b and c phase currents of the GSC
$V_{Sabc}$	a, b and c phase voltages of the grid
$P_{GSC}$	GSC active power
$Q_{GSC}$	GSC reactive power
$P_{MSC}$	MSC active power
$Q_{MSC}$	MSC reactive power

$P_{Cap}$	DC Link active power
$C$	capacitor
$V_{dqf}$	direct and quadrature phase voltage
$I_{dqf}$	direct and quadrature phase current
$U_s$	phase voltage
$V_{DC}$	DC Bus voltage
$L_f$	inductance of filter
$R_f$	resistor of filter

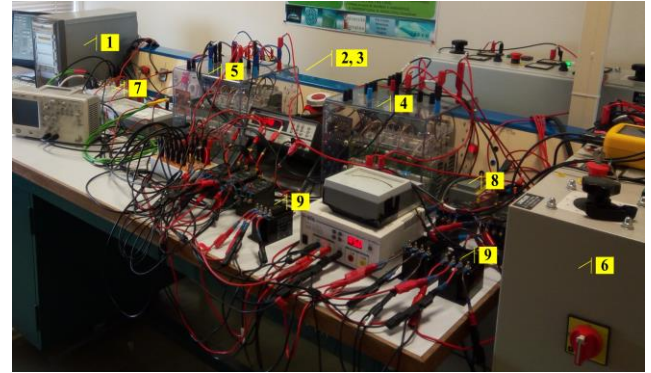
**Table 2.** Experimental bench nominal parameters

Settings	WT	DCM	SG
Nominal power	3 kW	3 kW	3 kW
Nominal speed	-	1,500 rpm	1,500 rpm
Nominal voltage	-	220 V	230 / 400 V

## APPENDIX

**Table 1.** WTE hardware component

Component	Name
1	PC
2	DCM
3	SG
4	Inverter (GSC)
5	Inverter (MSC)
6	Autotransformer
7	DS1104 interface
8	Filter
9	Sensors



**Figure 13.** Experimental laboratory setup